



Silicon Storage Technology, Inc.

Programming User's Manual

**Remote Controller
SST65P542R**



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1.0 INTRODUCTION

The SST65P542R is a member of SST's 8-bit, application-specific microcontroller family targeting IR remote controller applications.

The SST65P542R microcontroller provides high-functionality to infrared remote controller products. The device offers flexibility to store different remote control configurations for controlling multiple appliances. The configurations are either programmed at the factory during the manufacturing process or updated through a web download procedure using the serial interface.

Using the SuperFlash nonvolatile memory technology, the SST65P542R enhances the functionality of the conventional universal remote controller devices by integrating multiple functions of a remote controller system in a single chip solution. The built-in LED I/O ports can directly drive LED indicators. The IR transmitter port drives signals to the infrared transmitter, which, in turn, remotely controls appliances.

The SoftPartition flash memory architecture allows seamless partition of the program code, protocol tables, and user data in the small granularity of 128 Byte sectors. The small sector size and fast Erase/Write time greatly increase the time and power efficiency when altering the contents of the flash memory.

The embedded controller is designed and manufactured using SST's patented and proprietary SuperFlash EEPROM technology.

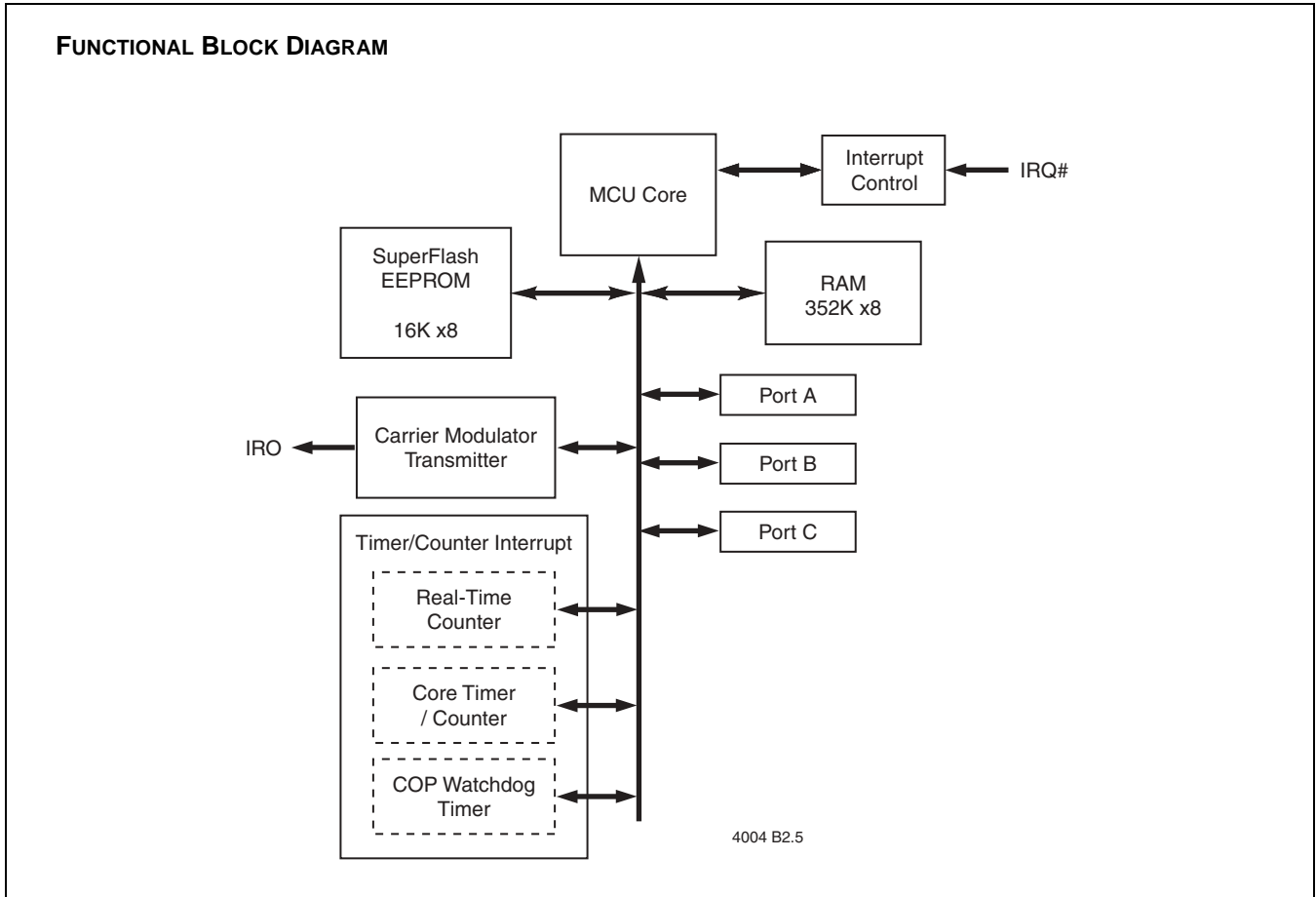
SST's highly reliable SuperFlash technology provides significant advantages over conventional flash memory technology. These advantages translate into significant overall cost savings and reliability benefits for customers.

PRODUCT FEATURES

- **8-bit MCU Core**
 - Enhanced 6502 Microprocessor Megacell
- **4 MHz Typical Oscillator Clock Frequency**
- **8 MHz maximum clock frequency**
- **16 KByte of user programmable flash memory**
- **352 Bytes SRAM**
- **IR Input Pin for Learning Mode**
- **Power-down Modes**
- **Carrier Modulator Transmitter**
 - Supports Baseband, Pulse Length Modulator (PLM), and Frequency Shift Keying (FSK)
- **Core Timer / Counter**
 - 14-stage multifunctional ripple counter
 - Includes timer overflow, POR, RTI, and CWT
- **General Registers:**
 - Accumulator (8-bit)
 - Index Register (8-bit)
- **Control registers:**
 - Program Counter (16-bit)
 - Stack Pointer (16-bit / 6 addressable bits)
 - Condition code register (8-bit)
- **Addressing modes supported:**
 1. Immediate
 2. Direct
 3. Extended
 4. Relative
 5. Indexed, no offset
 6. Indexed, 8-bit offset
 7. Indexed, 16-bit offset
 8. Bit set/clear
 9. Bit test and branch
 10. Inherent
- **Data types supported:**
 1. Bit data (manipulation instructions)
 2. Byte data



2.0 BLOCK DIAGRAM





3.0 PIN ASSIGNMENTS

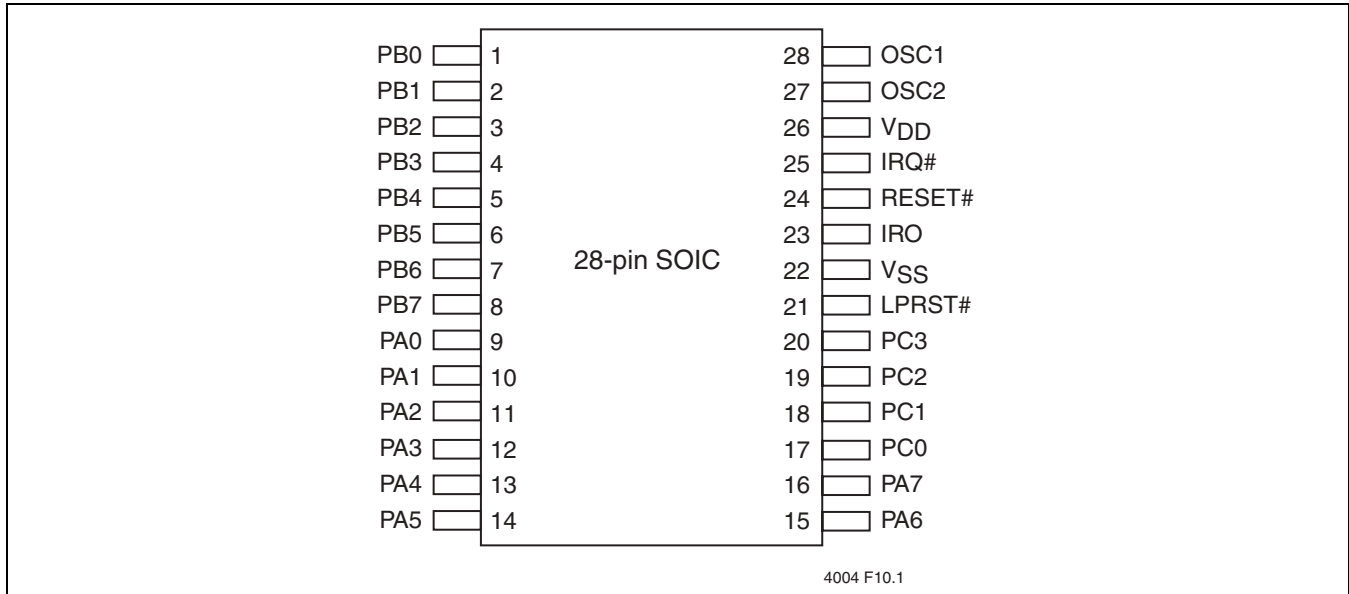


FIGURE 3-1: PIN ASSIGNMENTS FOR 28-PIN SOIC

TABLE 3-1: PIN DESCRIPTIONS

Pins	Symbol	Type ¹	Name and Functions
16-9	PA[7:0]	I/O ¹	Port A: The state of any pin in Port A is software programmable and every line is configured as an input during any external reset.
8-1	PB[7:0]	I/O with internal pull-ups	Port B: The state of any pin in Port B is software programmable and every line is configured as an input during any external reset. Each I/O line contains a programmable interrupt/pull-up for keyscan. PB[7] is used as a serial interface data line when the serial interface is enabled.
20-17	PC[3:0]	I/O	Port C: Every pin in Port C is a high-current pin and its state is software programmable. All lines are configured as inputs during any external reset.
23	IRO	O	IRO: Suitable for driving IR LED biasing logic, the IRO pin is the high-current source and sink output of the carrier modulator transmitter subsystem. Default state is low after any external reset.
21	LPRST#	I	Low-Power Reset: An active-low pin, LPRST# function sets MCU to low-power reset mode. The MCU, once in low-power reset mode, is held in reset with all processor clocks and crystal oscillator halted. An internal Schmitt trigger is included in the LPRST# pin to improve noise immunity.
24	RESET#	I	Reset: By setting the RESET# pin low the MCU is reset to a default state. An internal Schmitt trigger is included in the RESET# pin to improve noise immunity.
28	OSC1	I	Oscillator 1,2: These 2 pins interface with external oscillator circuits. A crystal resonator, a ceramic resonator, or an external clock signal can be used.
27	OSC2	O	
25	IRQ#	I	Interrupt Request: The IRQ# is negative edge-sensitive triggered. An internal Schmitt trigger is included in the IRQ# pin to improve noise immunity.
26	V _{DD}	I	Power Supply: Supply Voltage
22	V _{SS}	I	Ground: Circuit ground. (0V reference)

1. I = Input
O = Output



4.0 MEMORY ORGANIZATION

The SST65P542R has a total of 64 KByte of addressable memory. A memory map is shown in Figure 4-1. The memory consists of 32 Bytes of I/O registers, 352 Bytes of SRAM, 16 KByte of user flash memory, and 128 Bytes of user vectors.

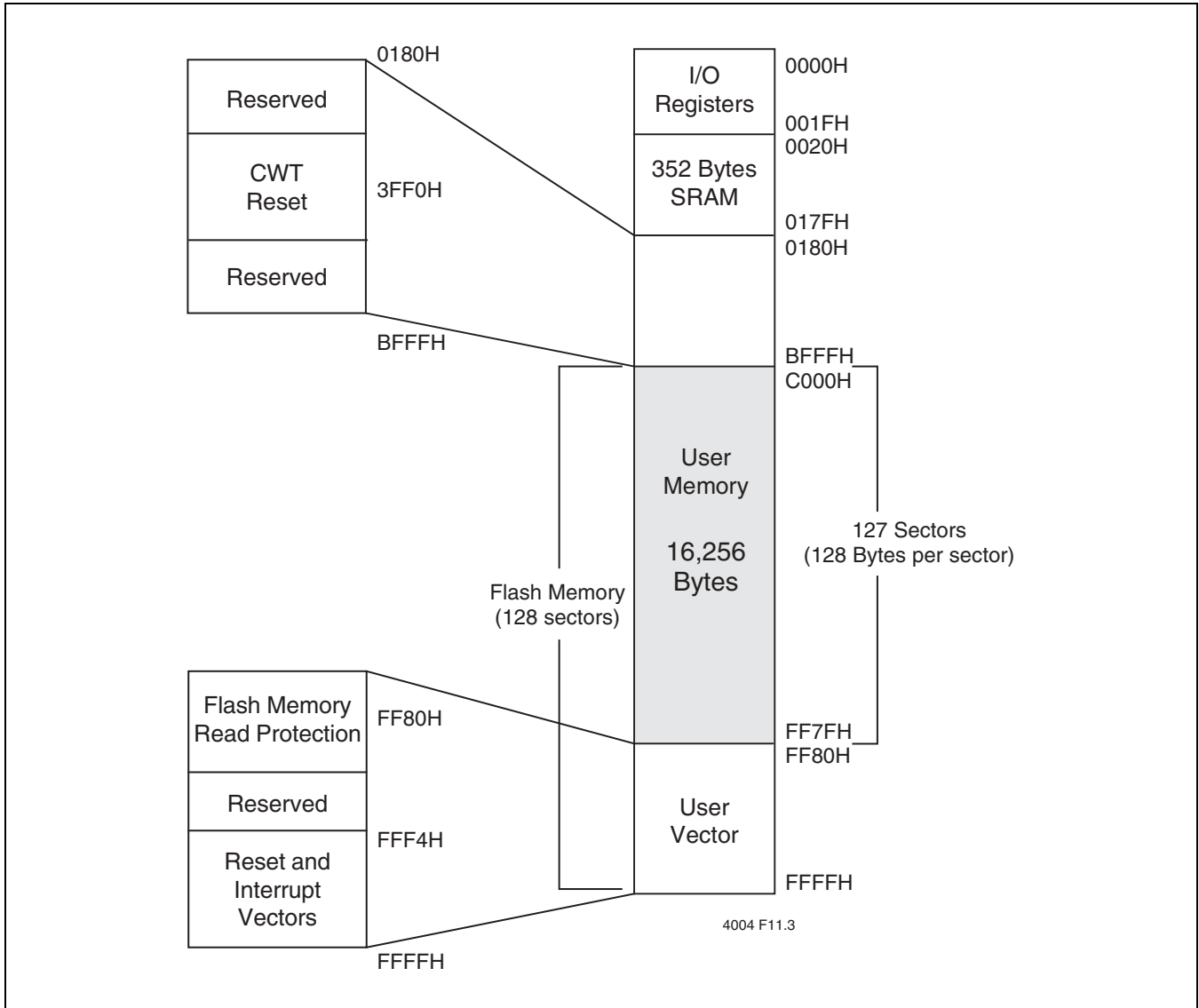


FIGURE 4-1: MEMORY MAP



5.0 MCU CORE AND INSTRUCTION SET

This section provides a description of the MCU core registers, the instruction set and the addressing modes.

5.1 Registers and Control Bit Assignments

The MCU contains five registers, as shown in the programming model of Figure 5-1. The interrupt stacking order is shown in Figure 5-2.

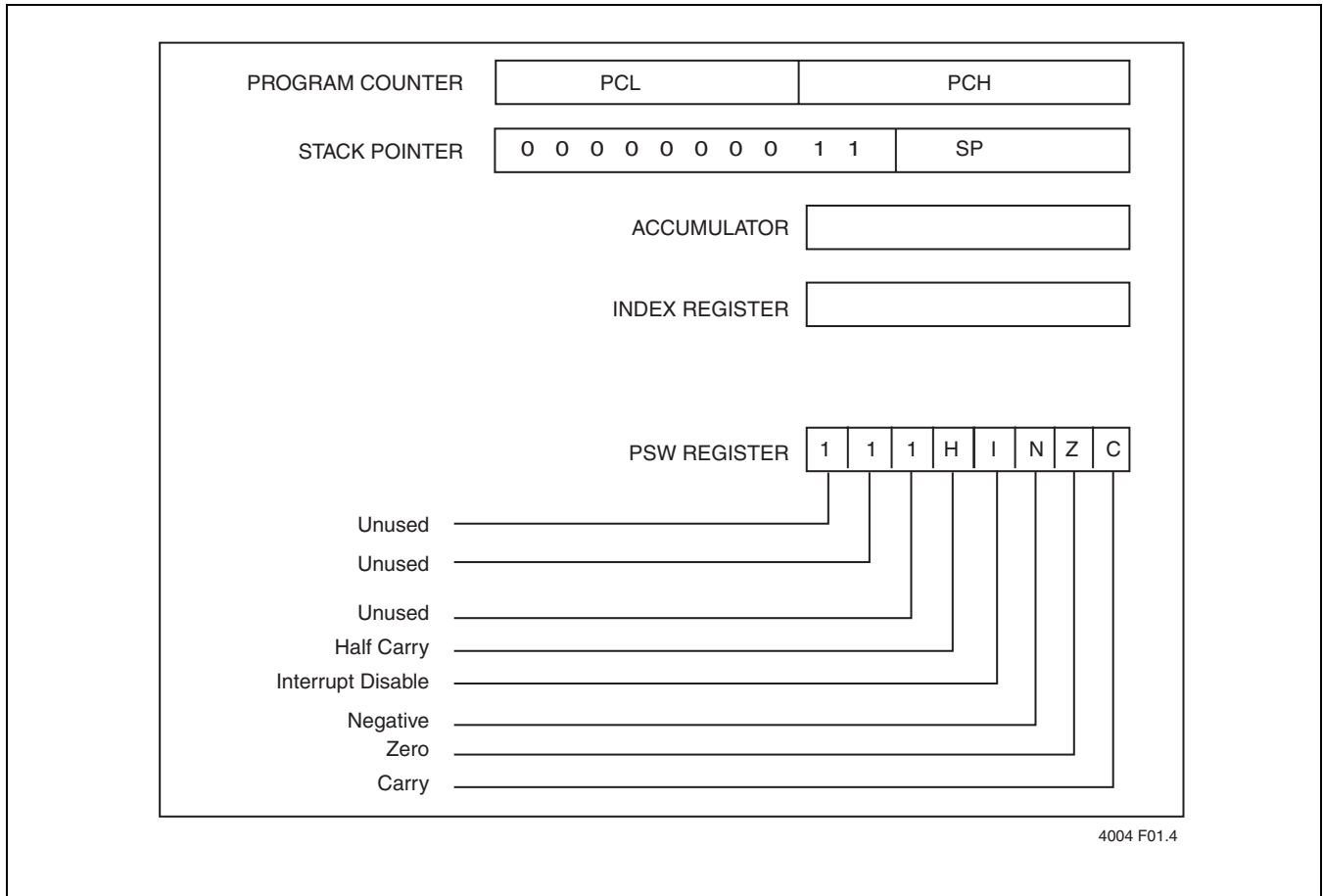


FIGURE 5-1: PROGRAMMING MODEL

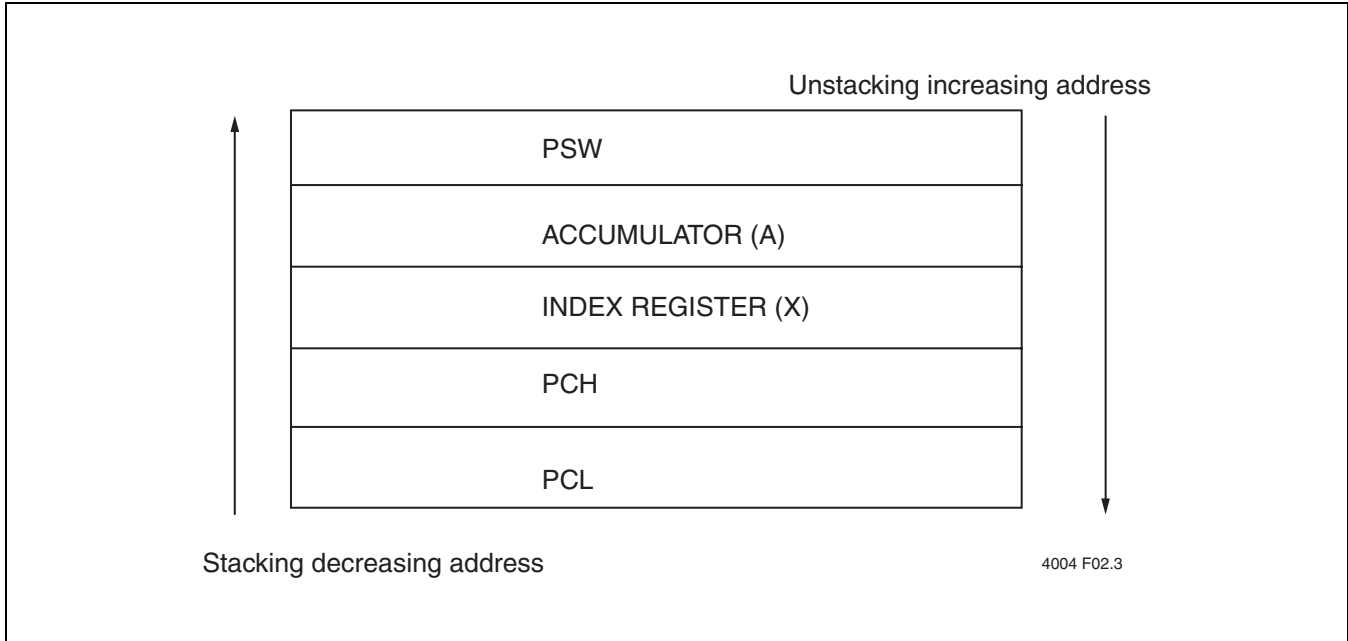


FIGURE 5-2: STACKING ORDER

Stacking decreases memory address and unstacking (Return) increases memory address.

5.1.1 Accumulator (A)

The accumulator is a general purpose 8-bit register used to hold operands and results of an arithmetic calculation or data manipulations.

5.1.2 Index Register (X)

The index register is an 8-bit register, which can contain the indexed addressing value used to create an effective address. The index register may also be used as a temporary storage area.

5.1.3 Program Counter (PC)

The program counter is a 16-bit register, which contains the address of the next byte to be fetched.

5.1.4 Stack Pointer (SP)

The stack pointer is a 16-bit register, which contains the address of the next free location on the stack. During an MCU reset or the reset stack pointer (RSP) instruction, the stack pointer is set to location 00FFH. The stack pointer is then decremented as data is pushed onto the stack and incremented as data is pulled from the stack. When accessing memory, the 8 most significant bits are permanently set to 00H. These eight bits are appended to the lower 8 significant register bits to produce an address within the range of 00C0H to 00FFH. Subroutines and interrupts may use up to 256 (decimal) locations. If 64 locations are exceeded, i.e. if stack pointer is pointing to 00C0H and stacking operation carried out, the stack pointer wraps around to 00FFH and overwrites the previously stored information. A subroutine call occupies two locations on the stack; an interrupt uses five locations.



5.1.5 Processor Status Word (PSW)

The PSW is a 5-bit register. These bits can be individually tested by a program, and specific actions can be taken as a result of their state. Each bit is explained in the following paragraphs.

5.1.5.1 Half Carry (H)

This bit is set during ADD and ADC operations to indicate that a carry has occurred between bits 3 and 4 of the accumulator during an ADD or ADC operation.

5.1.5.2 Interrupt (I)

When this bit is set all maskable interrupts are masked. If an interrupt occurs while this bit is set, the interrupt is latched and remains pending until the interrupt bit is cleared. After any reset, the interrupt mask is set and can be cleared by software instruction (CLI).

5.1.5.3 Negative (N)

When set, this bit indicates that the result of the last arithmetic, logical, or data manipulation was negative.

5.1.5.4 Zero (Z)

When set, this bit indicates that the result of the last arithmetic, logical, or data manipulation was zero.

5.1.5.5 Carry/Borrow (C)

When set, this bit indicates that a carry or borrow out of the arithmetic logical unit (ALU) occurred during the last arithmetic operation. This bit is also affected during bit test and branch instructions and during shifts and rotates.

5.2 Addressing Modes

Ten different addressing modes provide programmers with the flexibility to optimize their code for all situations. The various indexed addressing modes make it possible to locate data tables, code conversion tables and scaling tables anywhere in the memory space. Short indexed accesses are single byte instructions; the longest instructions (three bytes) enable access to tables throughout memory. Short absolute (direct) and long absolute (extended) addressing are also included. One or two byte direct addressing instructions access all data bytes in most applications.

Extended addressing permits jump instructions to reach all memory locations. The term 'effective address' (EA) is used in describing the various addressing modes. The effective address is defined as the address from which the argument for an instruction is fetched or stored. The ten addressing modes of the processor are described below. Parentheses are used to indicate 'contents of' the location or register referred to. For example, (PC) indicates the contents of the location pointed to by the PC (program counter). An arrow indicates 'is replaced by' and a colon indicates concatenation of two bytes.

5.2.1 Inherent (INH)

In the inherent addressing mode, all the information necessary to execute the instruction is contained in the opcode. Operations specifying only the index register or accumulator, as well as the control instruction, with no other arguments are included in this mode. These instructions are one byte long.

5.2.2 Immediate (IMM)

In the immediate addressing mode, the operand is contained in the byte immediately following the opcode.

$$EA = PC+1; PC \leftarrow PC+2$$



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5.2.3 Direct (DIR)

In the direct addressing mode, the effective address of the argument is contained in a single byte following the opcode byte. Direct addressing allows the user to directly address the lowest 256 bytes in memory with a single two-byte instruction.

$$EA = (PC+1); PC \leftarrow PC+2$$
$$\text{Address bus high byte} \leftarrow 0, \text{Address bus low byte} \leftarrow (PC+1)$$

5.2.4 Extended (EXT)

In the extended addressing mode, the effective address of the argument is contained in the two bytes following the opcode byte. Instructions with extended addressing mode are capable of referencing arguments anywhere in memory with a single three-byte instruction.

$$EA = (PC+1):(PC+2); PC \leftarrow PC+3$$
$$\text{Address bus high byte} \leftarrow (PC+1); \text{Address bus low byte} \leftarrow (PC+2)$$

5.2.5 Indexed, No Offset (IX)

In the indexed, no offset addressing mode, the effective address of the argument is contained in the 8-bit index register. This addressing mode can access the first 256 memory locations. These instructions are only one byte long. This mode is often used to move a pointer through a table or to hold the address of a frequently referenced RAM or I/O location.

$$EA = (X); PC \leftarrow PC+1$$
$$\text{Address bus high byte} \leftarrow 0; \text{Address bus low byte} \leftarrow (X)$$

5.2.6 Indexed, 8-bit Offset (IX1)

In the indexed, 8-bit offset addressing mode, the effective address is the sum of the contents of the unsigned 8-bit index register and the unsigned byte following the opcode. Therefore the operand can be located anywhere within the lowest 511 memory locations. This addressing mode is useful for selecting the m^{th} element in an n element table.

$$EA = (X)+(PC+1); PC \leftarrow PC+2$$
$$\text{Address bus high byte} \leftarrow K; \text{Address bus low byte} \leftarrow (X)+(PC+1)$$

where K = the carry from the addition of (X) and $(PC+1)$

5.2.7 Indexed, 16-bit Offset (IX2)

In the indexed, 16-bit offset addressing mode, the effective address is the sum of the contents of the unsigned 8-bit index register and the two unsigned bytes following the opcode. This address mode can be used in a manner similar to indexed, 8-bit offset except that this three-byte instruction allows tables to be anywhere in memory. As with direct and extended addressing, the assembler determines the shortest form of indexed addressing.

$$EA = (X)+[(PC+1):(PC+2)]; PC \leftarrow PC+3$$
$$\text{Address bus high byte} \leftarrow (PC+1)+K; \text{Address bus low byte} \leftarrow (X)+(PC+2)$$

where K = the carry from the addition of (X) and $(PC+2)$



5.2.8 Relative (REL)

The relative addressing mode is only used in branch instructions. In relative addressing, the contents of the 8-bit signed byte (the offset) following the opcode is sign-extended and added to the PC if, and only if, the branch conditions are true. Otherwise, control proceeds to the next instruction. The span of relative addressing is from -126 to +129 from the opcode address. The programmer need not calculate the offset when using the assembler, since it calculates the proper offset and checks to see that it is within the span of the branch.

$$\text{EA} = \text{PC}+2+(\text{PC}+1); \text{PC} \leftarrow \text{EA} \text{ if branch taken};$$
$$\text{otherwise PC} \leftarrow \text{PC}+2$$

5.2.9 Bit Set/Clear (BSC)

In the bit set/clear addressing mode, the bit to be set or cleared is part of the opcode. The byte following the opcode specifies the address of the byte in which the specified bit is to be set or cleared. Any read/write bit in the first 256 locations of memory can be selectively set or cleared with a single two-byte instruction.

$$\text{EA} = (\text{PC}+1); \text{PC} \leftarrow \text{PC}+2$$
$$\text{Address bus high byte} \leftarrow 0; \text{Address bus low byte} \leftarrow (\text{PC}+1)$$

5.2.10 Bit test and branch (BTB)

The bit test and branch addressing mode is a combination of direct addressing and relative addressing mode. The bit to be tested and its condition (set or clear) is included in the opcode. The address of the byte to be tested is in the single byte immediately following the opcode byte (EA1). The signed relative 8-bit offset in the third byte is sign-extended and added to the PC if the specified bit is set or cleared in the specified memory location. This single three-byte instruction allows the program to branch based on the condition of any readable bit in the first 256 locations of memory. The span of branch is from -125 to +130 from the opcode address. The state of the tested bit is also transferred to the carry bit of the condition code register.

$$\text{EA1} = (\text{PC}+1); \text{PC} \leftarrow \text{PC}+2$$
$$\text{Address bus high byte} \rightarrow 0; \text{Address bus low byte} \rightarrow (\text{PC}+1)$$
$$\text{EA2} = \text{PC}+3+(\text{PC}+2); \text{PC} \leftarrow \text{EA2} \text{ if branch taken};$$
$$\text{otherwise PC} \leftarrow \text{PC}+3$$



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5.3 Instruction Set

Table 5-1 summarizes the MCU instruction set. A description of the instructions and an explanation of abbreviations follows the table on page 18.

TABLE 5-1: MCU INSTRUCTION TABLE (1 OF 5)

Mnemonic	Explanation	Address Mode (n)	Hex Opcode	Number of Machine Cycles	Number of Bytes	Flags Affected
ADC	Add memory to accumulator with carry $A+M+C \rightarrow A$	IMM	A9	2	2	H – N Z C
		DIR	B9	3	2	
		EXT	C9	4	3	
		IX	F9	3	1	
		IX1	E9	4	2	
		IX2	D9	5	3	
ADD	Add memory to accumulator $A+M \rightarrow A$	IMM	AB	2	2	H – N Z C
		DIR	BB	3	2	
		EXT	CB	4	3	
		IX	FB	3	1	
		IX1	EB	4	2	
		IX2	DB	5	3	
AND	“AND” memory with accumulator $A\&M \rightarrow A$	IMM	A4	2	2	-- N Z –
		DIR	B4	3	2	
		EXT	C4	4	3	
		IX	F4	3	1	
		IX1	E4	4	2	
		IX2	D4	5	3	
ASR	Shift right one bit (accumulator or memory) $b0 \rightarrow C$ $b7$ held constant	INH (A)	47	3	1	-- N Z C
		INH (X)	57	3	1	
		DIR	37	5	2	
		IX	77	5	1	
		IX1	67	6	2	
ASL	(same as LSL) Shift left one bit (accumulator or memory) $b7 \rightarrow C$ $0 \rightarrow b0$	INH (A)	48	3	1	-- N Z C
		INH (X)	58	3	1	
		DIR	38	5	2	
		IX	78	5	1	
		IX1	68	6	2	
BCC	Branch on carry clear Branch on $C = 0$	REL	24	3	2	-----
BCLR	Clear bit n	BSC	$11 + 2n$	5	2	-----
BCS	Branch on carry set Branch on $C = 1$	REL	25	3	2	-----
BEQ	Branch on result zero Branch on $Z = 1$	REL	27	3	2	-----
BHCC	Branch if half carry clear Branch on $H = 0$	REL	28	3	2	-----
BHCS	Branch if half carry set Branch on $H = 1$	REL	29	3	2	-----
BHI	Branch if higher Branch if accumulator is higher than memory (unsigned)	REL	22	3	2	-----



TABLE 5-1: MCU INSTRUCTION TABLE (CONTINUED) (2 OF 5)

Mnemonic	Explanation	Address Mode (n)	Hex Opcode	Number of Machine Cycles	Number of Bytes	Flags Affected
BHS	Branch if higher or same Branch if accumulator is higher or same as memory (C = 0)	REL	24	3	2	-----
BIH	Branch if interrupt line is high	REL	2F	3	2	-----
BIL	Branch if interrupt line is low	REL	2E	3	2	-----
BIT	Tests bits in memory: A^M (logical compare)	IMM	A5	2	2	--NZ-
		DIR	B5	3	2	
		EXT	C5	4	3	
		IX	F5	3	1	
		IX1	E5	4	2	
		IX2	D5	5	3	
BLO	Branch if lower Branch if accumulator is lower Branch on	REL	25	3	2	-----
BLS	Branch if low or same Branch if accumulator is lower than or equal to memory	REL	23	3	2	-----
BMC	Branch if interrupt mask bit is clear	REL	2C	3	2	-----
BMI	Branch if minus branch on N = 1	REL	2B	3	2	-----
BMS	Branch if interrupt mask bit is set Branch on I = 1	REL	2D	3	2	-----
BNE	Branch if not equal Branch on Z = 0	REL	26	3	2	-----
BPL	Branch if plus Branch on N = 0	REL	2A	3	2	-----
BRA	Branch always	REL	20	3	2	-----
BRN	Branch never	REL	21	3	2	-----
BRCLR	Branch if bit n is clear	BTB	1 + 2n	5	3	----C
BRSET	Branch if bit n is set	BTB	2n	5	3	----C
BSET	Set bit n	BSC	10 + 2n	5	2	-----
BSR	Branch to subroutine	REL	AD	6	2	-----
CLC	Clear carry flag 0 →C	INH	98	2	1	----0
CLI	Clear interrupt mask bit 0 →I	INH	9A	2	1	-0---
CLR	Clear	INH (A)	4F	3	1	--01-
		INH (X)	5F	3	1	
		DIR	3F	5	2	
		IX	7F	5	1	
		IX1	6F	6	2	



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TABLE 5-1: MCU INSTRUCTION TABLE (CONTINUED) (3 OF 5)

Mnemonic	Explanation	Address Mode (n)	Hex Opcode	Number of Machine Cycles	Number of Bytes	Flags Affected
CMP	Arithmetic compare memory and accumulator (unsigned) A - M	IMM	A1	2	2	--NZC
		DIR	B1	3	2	
		EXT	C1	4	3	
		IX	F1	3	1	
		IX1	E1	4	2	
		IX2	D1	5	3	
COM	Component (one's complement)	INH (A)	43	3	1	---NZ1
		INH (X)	53	3	1	
		DIR	33	5	2	
		IX	73	5	1	
		IX1	63	6	2	
CPX	Arithmetic compare memory and index X (unsigned) X-M	IMM	A3	2	2	--NZC
		DIR	B3	3	2	
		EXT	C3	4	3	
		IX	F3	3	1	
		IX1	E3	4	2	
		IX2	D3	5	3	
DEC	Decrement by one	INH (A)	4A	3	1	---NZ-
		INH (X)	5A	3	1	
		DIR	3A	5	2	
		IX	7A	5	1	
		IX1	6A	6	2	
EOR	"Exclusive or" memory with accumulator $A \wedge M \rightarrow A$	IMM	A8	2	2	---NZ-
		DIR	B8	3	2	
		EXT	C8	4	3	
		IX	F8	3	1	
		IX1	E8	4	2	
		IX2	D8	5	3	
INC	Increment by one	INH (A)	4C	3	1	---NZ-
		INH (X)	5C	3	1	
		DIR	3C	5	2	
		IX	7C	5	1	
		IX1	6C	6	2	
JMP	Jump to new location (PC + 1) \rightarrow PCL (PC + 2) \rightarrow PCH	DIR	BC	2	2	-----
		EXT	CC	3	3	
		IX	FC	2	1	
		IX1	EC	3	2	
		IX2	DC	4	3	
JSR	Jump to new location saving return address PC + 2 \downarrow (PC + 1) \rightarrow PCL (PC + 2) \rightarrow PCH	DIR	BD	5	2	-----
		EXT	CD	6	3	
		IX	FD	5	1	
		IX1	ED	6	2	
		IX2	DD	7	3	



TABLE 5-1: MCU INSTRUCTION TABLE (CONTINUED) (4 OF 5)

Mnemonic	Explanation	Address Mode (n)	Hex Opcode	Number of Machine Cycles	Number of Bytes	Flags Affected
LDA	Load accumulator with memory M →A	IMM	A6	2	2	--NZ-
		DIR	B6	3	2	
		EXT	C6	4	3	
		IX	F6	3	1	
		IX1	E6	4	2	
		IX2	D6	5	3	
LDX	Load index X with memory M →X	IMM	AE	2	2	--NZ-
		DIR	BE	3	2	
		EXT	CE	4	3	
		IX	FE	3	1	
		IX1	EE	4	2	
		IX2	DE	5	3	
LSL	(same as ASL) Shift left one bit (accumulator or memory) b7 →C 0 →b0	INH (A)	48	3	1	--NZC
		INH (X)	58	3	1	
		DIR	38	5	2	
		IX	78	5	1	
		IX1	68	6	2	
LSR	Shift right one bit (memory or accumulator) b0 →C 0 →b7	INH (A)	44	3	1	--0ZC
		INH (X)	54	3	1	
		DIR	34	5	2	
		IX	74	5	1	
		IX1	64	6	2	
MUL	Multiplication X * A →X: A	INH (A)	42	11	1	0---0
NEG	Negate (Two's complement)	INH (A)	40	3	1	--NZC
		INH (X)	50	3	1	
		DIR	30	5	2	
		IX	70	5	1	
		IX1	60	6	2	
NOP	No operation	INH	9D	2	1	-----
ORA	"OR" memory with accumulator A M →A	IMM	AA	2	2	--NZ-
		DIR	BA	3	2	
		EXT	CA	4	3	
		IX	FA	3	1	
		IX1	EA	4	2	
		IX2	DA	5	3	
ROL	Rotate one bit left through carry (memory or accumulator)	INH (A)	49	3	1	--NZC
		INH (X)	59	3	1	
		DIR	39	5	2	
		IX	79	5	1	
		IX1	69	6	2	
ROR	Rotate one bit right through carry (memory or accumulator)	INH (A)	46	3	1	--NZC
		INH (X)	56	3	1	
		DIR	36	5	2	
		IX	76	5	1	
		IX1	66	6	2	
RSP	Reset stack pointer	INH	9C	2	1	-----
RTI	Return from interrupt PC; P	INH	80	9	1	?????



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TABLE 5-1: MCU INSTRUCTION TABLE (CONTINUED) (5 OF 5)

Mnemonic	Explanation	Address Mode (n)	Hex Opcode	Number of Machine Cycles	Number of Bytes	Flags Affected
RTS	Return from subroutine PC ↑ ; PC + 1 → PC	INH	81	6	1	-----
SBC	Subtract memory from accumulator with borrow A-M-C → A	IMM	A2	2	2	--NZC
		DIR	B2	3	2	
		EXT	C2	4	3	
		IX	F2	3	1	
		IX1	E2	4	2	
IX2	D2	5	3			
SEC	Set carry flag 1 → C	INH	99	2	1	-----1
SEI	Set interrupt mask bit I → 1	INH	9B	2	1	-1----
STA	Store accumulator in memory A → M	DIR	B7	4	2	--NZ-
		EXT	C7	5	3	
		IX	F7	4	1	
		IX1	E7	5	2	
		IX2	D7	6	3	
STX	Store index X in memory X → M	DIR	BF	4	2	--NZ-
		EXT	CF	5	3	
		IX	FF	4	1	
		IX1	EF	5	2	
		IX2	DF	6	3	
SUB	Subtract memory	IMM	A0	2	2	--NZC
		DIR	B0	3	2	
		EXT	C0	4	3	
		IX	F0	3	1	
		IX1	E0	4	2	
		IX2	D0	5	3	
SWI	Software interrupt	INH	83	10	1	-1----
TAX	Transfer accumulator to index X A → X	INH	97	2	1	-----
TST	Test for negative or zero	INH (A)	4D	3	1	--NZ-
		INH (X)	5D	3	1	
		DIR	3D	4	2	
		IX	7D	4	1	
		IX1	6D	5	2	
TXA	Transfer index X to accumulator X → A	INH	9F	2	1	-----

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Description

The following is the description of each instruction and the operation during the execution of each instruction.

The key for MCU instructions is as follows:

The first three letters are the opcode the actual mnemonic of the instruction. The possible addressing modes are indicated by the letters following the opcode and they are as follows:

IMM:	immediate addressing
INH (A):	inherent addressing with respect to Accumulator
INH (X):	inherent addressing with respect to Index Register
DIR:	direct addressing
EXT:	extended addressing
IX:	indexed addressing (no offset)
IX1:	indexed addressing with one byte offset
IX2:	indexed addressing with two byte offset
BSC:	bit set / clear
BTB:	bit test and branch
REL:	relative addressing

The following abbreviations are used besides the ones used for addressing mode:

A	accumulator
C	carry flag
H	half carry flag
I	interrupt flag
M	memory
N	negative flag
PC	program counter
PCL	program counter lower byte
PCH	program counter higher byte
SP	stack pointer
X	index register
Z	zero flag
	OR function
&	AND function
^	Exclusive OR function
???	load PSW from stack
-	not affected

Machine cycle is two oscillator clock cycles.



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6.0 I/O REGISTERS DEFINITION

The 32 Bytes of I/O registers occupy address locations from 0000H to 001FH and include general purpose I/O pin registers, on-chip peripheral control registers, and SuperFlash Function Registers.

TABLE 6-1: I/O REGISTER DESCRIPTIONS

Address Location	Register Description
0000H	Port A Data Register
0001H	Port B Data Register
0002H	Port C Data Register
0003H	Reserved
0004H	Port A Data Direction Register
0005H	Port B Data Direction Register
0006H	Port C Data Direction Register
0007H	Reserved
0008H	Core Timer Control Status Register
0009H	Core Timer Counter Register
000AH	Port B Interrupt Control Register
000BH	SuperFlash Function Register (SFFR)
000CH	Port B Pull-up Control Register
000DH	COP Watchdog Timer Control Register (CWTC)
000EH	Serial Interface Control Register (SICON_TR)
000FH	Serial Interface Control Register (SICON_LSBF)
0010H	Carrier Generator High Data Register1 (CHR1)
0011H	Carrier Generator Low Data Register1 (CLR1)
0012H	Carrier Generator High Data Register2 (CHR2)
0013H	Carrier Generator Low Data Register2 (CLR2)
0014H	Modulator Control and Status Register (MCSR)
0015H	Modulator Data Register1 (MDR1)
0016H	Modulator Data Register2 (MDR2)
0017H	Modulator Data Register3 (MDR3)
0018H	Power Save Control Register (PSCR)
0019H	Serial Interface Control Register (SICON_SI)
001AH	Serial Interface Data Register (SIDAT)
001BH	Serial Interface Status Register (SISTA)
001CH	Serial Interface Baud-Rate Register (SIBDR)
001DH	Serial Interface Control Register (SICON_AP)
001EH	Serial Interface Control Register (SICON_ENSI)
001FH	IR Input Control Register

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TABLE 6-2: BIT DEFINITIONS OF I/O REGISTERS (1 OF 2)

Addr	Register Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
0000H	Port A Data Register	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
0001H	Port B Data Register	PB7	PB6	PB5	PB4	PB3	PB2	PB1	PB0
0002H	Port C Data Register	X	X	X	X	PC3	PC2	PC1	PC0
0003H	Reserved	X	X	X	X	X	X	X	X
0004H	Port A Data Direction Register	DDRA7	DDRA6	DDRA5	DDRA4	DDRA3	DDRA2	DDRA1	DDRA0
0005H	Port B Data Direction Register	DDRB7	DDRB6	DDRB5	DDRB4	DDRB3	DDRB2	DDRB1	DDRB0
0006H	Port C Data Direction Register	X	X	X	X	DDRC3	DDRC2	DDRC1	DDRC0
0007H	Reserved	X	X	X	X	X	X	X	X
0008H	Core Timer Control Status Register	CTOF	RTIF	TOFE	RTIE	TOFC	TRFC	RT1	RT0
0009H	Core Timer Counter Register	X	X	X	X	X	X	X	X
000AH	Port B Interrupt Control Register	INPRB7	INPRB6	INPRB5	INPRB4	INPRB3	INPRB2	INPRB1	INPRB0
000BH	SuperFlash Function Register	PREN	MEREN	SEREN	X	PROG	MERA	SERA	X
000CH	Port B Pull-up Control Register	X	X	X	X	X	X	PU1	PU0
000DH	CWT Control Register	X	X	X	X	X	X	X	CWT_EN
000EH	Serial Interface Control Register	X	X	X	X	X	X	TR	X
000FH	Serial Interface Control Register	X	X	X	X	X	LSBF	X	X
0010H	Carrier Generator High Data Register (CHR1)	IROLN	CMT-POL	PH5	PH4	PH3	PH2	PH1	PH0
0011H	Carrier Generator Low Data Register (CLR1)	IROLP	X	PL5	PL4	PL3	PL2	PL1	PL0
0012H	Carrier Generator High Data Register 2(CHR2)	X	X	SH5	SH4	SH3	SH2	SH1	SH0
0013H	Carrier Generator Low Data Register 2(CLR2)	X	X	SL5	SL4	SL3	SL2	SL1	SL0
0014H	Modulator Control and Status Register (MCSR)	EOC	DIV2	EIMSK	EXSPC	BASE	MODE	EOCIE	MCGEN
0015H	Modulator Data Register 1(MDR1)	MB11	MB10	MB9	MB8	SB11	SB10	SB9	SB8
0016H	Modulator Data Register 2(MDR2)	MB7	MB6	MB5	MB4	MB3	MB2	MB1	MB0
0017H	Modulator Data Register 3(MDR3)	SB7	SB6	SB5	SB4	SB3	SB2	SB1	SB0
0018H	Power Saving Control Register (PSCR)	EN	X	X	X	X	X	STOP	IDL
0019H	Serial Interface Control Register	X	X	X	X	X	X	X	SI

X = Reserved (Recommended to write "0" to reserved bits for future compatibility)

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TABLE 6-2: BIT DEFINITIONS OF I/O REGISTERS (CONTINUED) (2 OF 2)

Addr	Register Name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0
001AH	Serial Interface Data Register	D7	D6	D5	D4	D3	D2	D1	D0
001BH	Serial Interface Status Register	X	X	S5	S4	S3	X	X	X
001CH	Serial Interface Baud-Rate Register	X	X	X	X	F3	F2	F1	F0
001DH	Serial Interface Control Register	X	X	X	X	AP	X	X	X
001EH	Serial Interface Control register	X	X	X	ENSI	X	X	X	X
001FH	IR Input Control Register	X	X	X	X	X	X	IREN	IRIN

X = Reserved (Recommended to write "0" to reserved bits for future compatibility)

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Please see Section 10.0 for Core Timer and Section 11.0 for CMT register definitions. All other register definitions are described in detail in the SST65P542R data sheet.



7.0 INTERRUPTS

The MCU has 6 sources of interrupts including reset, a software interrupt and 4 hardware interrupt lines. If more than one interrupt line is active, the one with highest priority will be serviced first. The interrupt priority from high to low is hardware reset, software interrupt, external interrupt, CMT interrupt, Core Timer interrupt, and Serial Interface interrupt. See Table 7-1 for Interrupt Address Vectors.

TABLE 7-1: MCU INTERRUPT ADDRESS VECTORS

Interrupt	Address	Description
Reset	FFFEH-FFFFH	Restart vector FFFEh higher byte, FFFFh lower byte
SWI	FFFCH-FFFDH	Software interrupt vector FFFCh higher byte, FFFDh lower byte
External/Port B	FFFAH-FFFBH	External/Port B interrupt vector FFFAh higher byte, FFFBh lower byte
CMT	FFF8H-FFF9H	CMT interrupt vector FFF8h higher byte, FFF9h lower byte
Core Timer	FFF6H-FFF7H	Core Timer interrupt vector FFF6h higher byte, FFF7h lower byte
Serial Interface	FFF4H-FFF5H	Serial Interface interrupt vector FFF4h higher byte, FFF5h lower byte

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All four interrupt lines are masked by the interrupt mask bit (I bit of the Process Status Word Register). The software interrupt is generated by SWI instruction similar to that of any hardware instruction except that it is not maskable (the execution of the SWI instruction is independent of the state of the mask bit).

All external interrupt lines are falling edge triggered and interrupts are always checked before fetching the next instruction.

If interrupt is recognized, before the Program counter jumps to one of the address vectors, the Program Counter, Index Register, Accumulator, and the Process Status Word Register are pushed on to the stack (see Figure 5-2).



8.0 RESETS AND CLOCKS

SST65P542R has two sources for external reset: LPRST# and RESET#.

After LPRST# switches from low to high, 4064 clock cycles are counted before the reset vector address appears on the internal address bus. RESET# immediately resets the MCU without counting the 4064 clock cycles.

Crystal oscillator clock is divided by two to arrive at internal processor and peripheral clock. Figure 8-1 shows RESET# and LPRST# timing diagram.

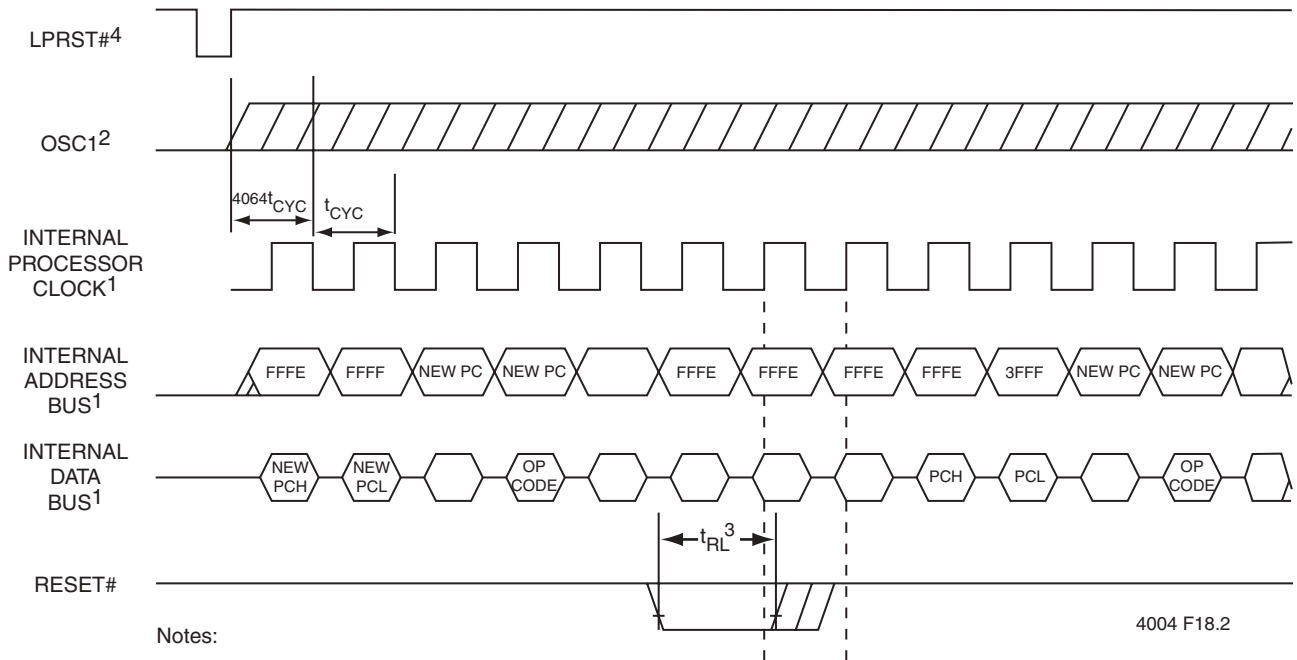


FIGURE 8-1: RESET# AND LPRST# TIMING DIAGRAM



9.0 POWER-DOWN MODES

SST65P542R offers two modes to reduce system power consumption.

9.1 STOP Mode

To enter the STOP Mode, write 01H to the Power Saving Control Register (PSCR - 0018H). Upon completion of the Write operation to the PSCR, the internal oscillator is turned off, halting all internal processing, including CMT and timer operations.

The microcontroller can be brought out of the STOP Mode by external/Port B interrupts, LPRST# or RESET#.

When external interrupt is asserted (either on IRQ# or Port B pins (when Port B interrupt is enabled)), interrupt will be serviced if mask bit (I bit) is clear, otherwise interrupt will not be serviced. MCU will resume operation to the next instruction byte following the STOP mode enabling Write operation.

The STOP bit will be set to 1 when the device has been brought out of STOP mode. The I bit will not be affected.

9.2 IDLE Mode

To enter IDLE Mode, write 02H to Power Saving Control Register (0018H). The IDLE Mode consumes more power than the STOP Mode. Upon completion of the Write operation to the PSCR, all modules remain active except MCU clock processing is suspended. Any interrupt or reset will cause the MCU to exit IDLE mode.

When an interrupt is asserted, interrupt will be serviced if mask bit (I bit) is clear, otherwise interrupt will not be serviced. MCU will resume operation to the next instruction byte following the IDLE mode enabling Write operation. The IDLE bit will be set to 1 when the device has been brought out of IDLE mode. The I bit will not be affected.



10.0 THE CORE TIMER

The core timer is a 14-stage, multifunctional ripple counter. Its features include Timer Overflow (TO), Power-On Reset (POR), Real-Time Interrupt (RTI), and COP Watchdog Timer (CWT).

The core timer operates as follows:

1. The internal peripheral clock is divided by four, driving an 8-bit ripple counter. At any time, the counter value can be read by accessing the Timer Counter Register (TCR) address 0009H.
2. At the last stage of the counter, a timer overflow is implemented. This gives a possible interrupt rate of the internal peripheral clock, $E/1024$.
3. After three more stages, the clock, RTI_{OUT} , now $E/4096$, drives the real-time interrupt circuit (RTI).
4. The RTI circuit has three divider stages with a 4:1 selector.
5. The output of the RTI circuit is divided by eight. This drives the CWT circuit.
6. The Timer Control and Status register at location 0008H contains the RTI rate selector bits and the RTI and CTOF enable bits and flags.

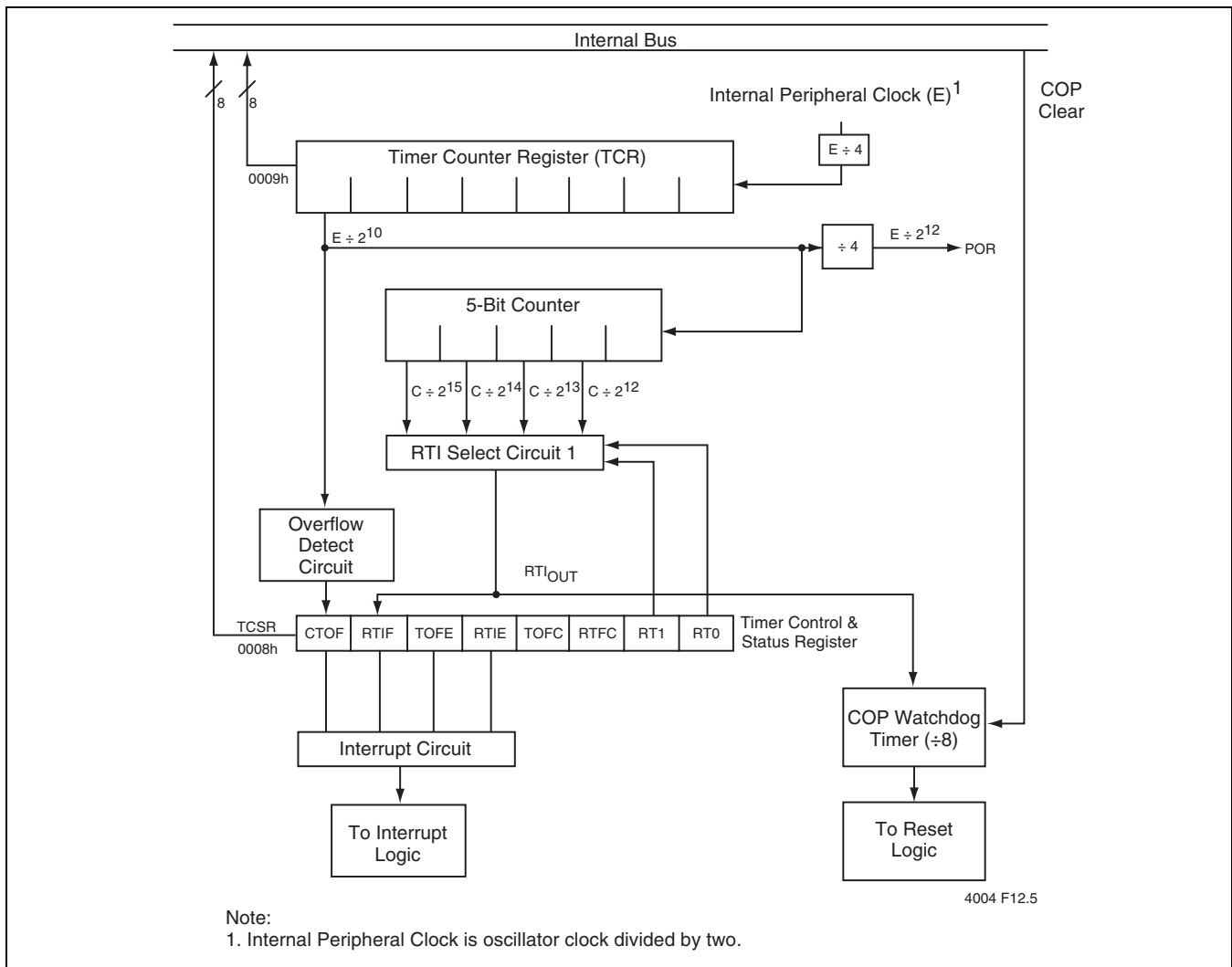


FIGURE 10-1: CORE TIMER BLOCK DIAGRAM



10.1 Computer Operating Properly Watchdog Timer Control Register (CWTC)

Writing to CWT Control register (000DH) can enable/disable core timer.

CWT Control Register

Address: 000DH

Bit 7	6	5	4	3	2	1	Bit 0
R	X	X	X	X	X	X	CWT_EN
Default							1

CWT_EN: '1' CWT disabled;'0' CWT enabled

10.2 Timer Control and Status Register (TCSR)

The TCSR includes the timer interrupt flag, the timer interrupt enable bits, and the real-time interrupt rate select bits. The table below shows the default value of each bit in the TCSR immediately after reset.

Timer Control and Status Register

Address: 0008H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	CTOF	RTIF	TOFE	RTIE	0	0	RT1	RT0
Write:	X	X			TOFC	RTFC		
Reset	0	0	0	0	0	0	1	1

10.2.1 Core Timer Overflow (CTOF)

CTOF is a read-only status bit set when the 8-bit ripple counter rolls over from FFH to 00H. Writing a logical "1" to TOFC bit clears the CTOF. Writing to CTOF has no affect. Reset clears CTOF.

10.2.2 Real-Time Interrupt Flag (RTIF)

The Real Time Interrupt circuit contains a 3-stage divider and a one-of-four choice selector. The input frequency to the RTI is (E/4096) and after three other divider stages allows a maximum interrupt period of 16 milliseconds at an internal peripheral clock rate of 2.048MHz. RTIF is a read-only status bit which is set when the output of the selected (one-of-four) divider stage goes active. Clearing the RTIF is done by writing a logical "1" to RTFC. Writing to RTIF has no affect. Reset clears the RTIF bit.

10.2.3 Timer Overflow Enable (TOFE)

When the TOFE bit is set, a MCU interrupt request is generated only if CTOF bit is set. Reset clears this bit.

10.2.4 Real-Time Interrupt Enable (RTIE)

When the RTIE bit is a set, a MCU interrupt request is generated only if RTIF bit is set. Reset clears this bit.

10.2.5 Timer Overflow Flag Clear (TOFC)

CTOF is cleared when logical "1" is written to TOFC. Writing a "0" to TOFC has no effect on CTOF. TOFC has a "0" as default.

10.2.6 Real-Time Interrupt Flag Clear (RTFC)

RTIF is cleared when RTFC bit is written to as "1". Writing a "0" to RTFC has no effect on RTIF. RTFC has "0" as default.



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10.2.7 Real-Time Interrupt Rate Select (RT1-RT0)

These two bits select any one-of-four taps from the real-time interrupt circuit stages. Please reference Table 10-1. Reset sets these two bits to 11, selecting the lowest periodic rate, and gives the maximum time to alter these bits if necessary. CWT should be cleared before changing RTI taps. If the selected tap is modified during a cycle in which the counter is switching, a RTIF could be missed or an additional RTIF could be generated.

TABLE 10-1: RTI AND CWT RATES AT 4.096 MHz OSCILLATOR

RTI Rate E = 2.048-MHz		RT1-RT0	Minimum CWT Rates E = 2.048-MHz		Maximum CWT Rates E = 2.048-MHz	
2 ms	$2^{12} \div E$	00	$(2^{15}-2^{12})/E$	14 ms	$(2^{15})/E$	16 ms
4 ms	$2^{13} \div E$	01	$(2^{16}-2^{13})/E$	28 ms	$(2^{16})/E$	32 ms
8 ms	$2^{14} \div E$	10	$(2^{17}-2^{14})/E$	56 ms	$(2^{17})/E$	64 ms
16 ms	$2^{15} \div E$	11	$(2^{18}-2^{15})/E$	112 ms	$(2^{18})/E$	128 ms

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10.3 Core Timer Counter Register (CTCR)

The TCR is a read-only register that contains the current value of the 8-bit ripple counter. This counter is clocked by divided-by-four peripheral clock (E/4) and can be used for various timing related functions, including a software-input capture. Extended time periods can be achieved using the Timer Overflow function to increment a temporary RAM storage location, there by simulating a 16-bit or larger counter.

Core Timer Counter Register (CTCR)

Address: 0009H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	D7	D6	D5	D4	D3	D2	D1	D0
Write:	X	X	X	X	X	X	X	X
Reset	0	0	0	0	0	0	1	1

10.4 COP Watchdog Timer (CWT) Reset

The CWT objective is to prevent the device to become stuck or locked-up.

The COP Watchdog Timer (CWT) function is achieved by using the output of the RTI circuit and further dividing by 8. The minimum reset rates are listed in Table 10-1.

If the CWT circuit times out, an internal reset is generated. This internal reset is equivalent to RESET# pin reset.

To clear the CWT, write 00H to address 3FF0H. When CWT is cleared, only the final divide-by-eight output of the RTI is cleared.

10.5 Timer During IDLE Mode

The MCU clock is stopped during IDLE mode, but the timer remains active. If interrupts are enabled, a timer interrupt will cause the processor to exit IDLE mode.

11.0 CARRIER MODULATOR TRANSMITTER (CMT)

The carrier modulator transmitter (CMT) module is tailored for the IR remote controller applications. This module is built by hardware with programmable ability for a wide variety of encoding schemes. The incorporated hardware can off-load MCU to perform lengthy time-consuming tasks associated with code generation. It's designed to handle most of the protocols. When a special protocol is needed, the CMT modulator can be disabled. A CMT register can be used to change the state of the infrared out pin (IRO) directly.

The CMT module consists of three blocks: carrier generator, modulator, and transmitter output. The block diagram is shown in Figure 11-1.

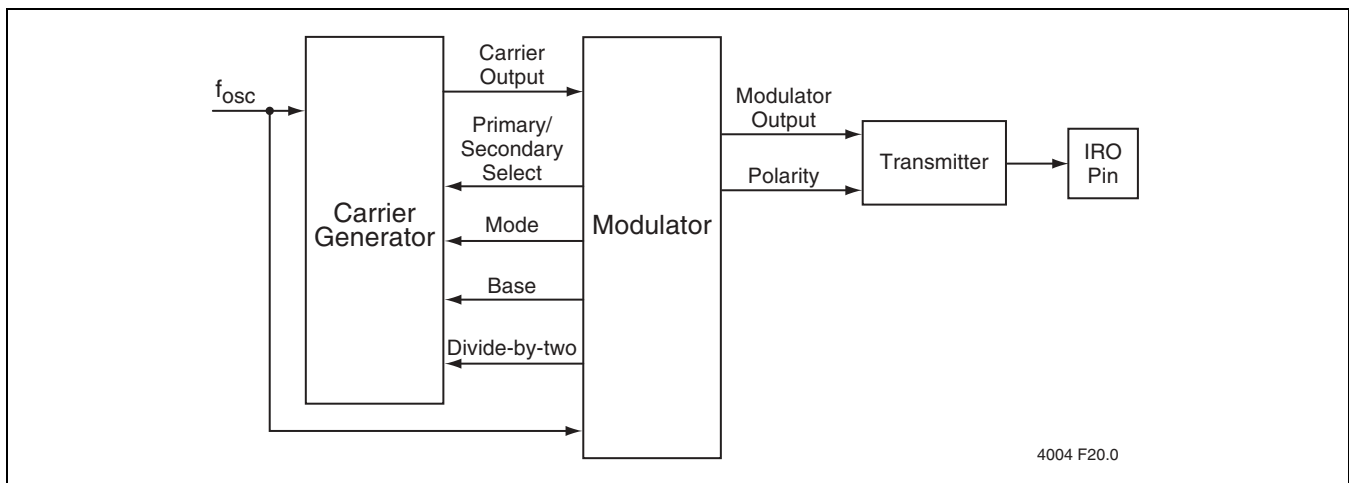


FIGURE 11-1: CARRIER MODULATOR TRANSMITTER MODULE BLOCK DIAGRAM

11.1 Carrier Generator

The carrier generator has a resolution of 500ns time steps with a 2MHz f_{osc} . The high and low times of the carrier signal can be programmed by user independently to determine both period and duty cycle of carrier signal. The period of carrier signal can be from 1 us (1MHz) to 64 us (15.6KHz) in 500ns increments. The duty cycle resolution is depend on the number of counts required to complete the carrier period. These counts are split between high and low times of the carrier signal. The longer the carrier signal period (the lower carrier signal frequency) the higher the resolution (as a percentage of the total period) of carrier signal duty cycle.

In carrier generator block, there are two sets of high and low times, i.e., two set of carrier signal period (frequency) can be selected. In normal mode (subsequently referred to as time mode), just one set will be used. In FSK (frequency shift key) mode, the modulator selects the two sets of carrier frequency base on the data to be modulated to generate dual frequency FSK protocols without MCU intervention. When the BASE bit in the modulator control and status register (MCSR) is set, the carrier output to the modulator is held high continuously to allow for the generation of baseband protocols.

To enable carrier generator clocks, the MCGEN bit in the MCSR must be set and the BASE bit in the MCSR must be cleared. The block diagram is shown below:

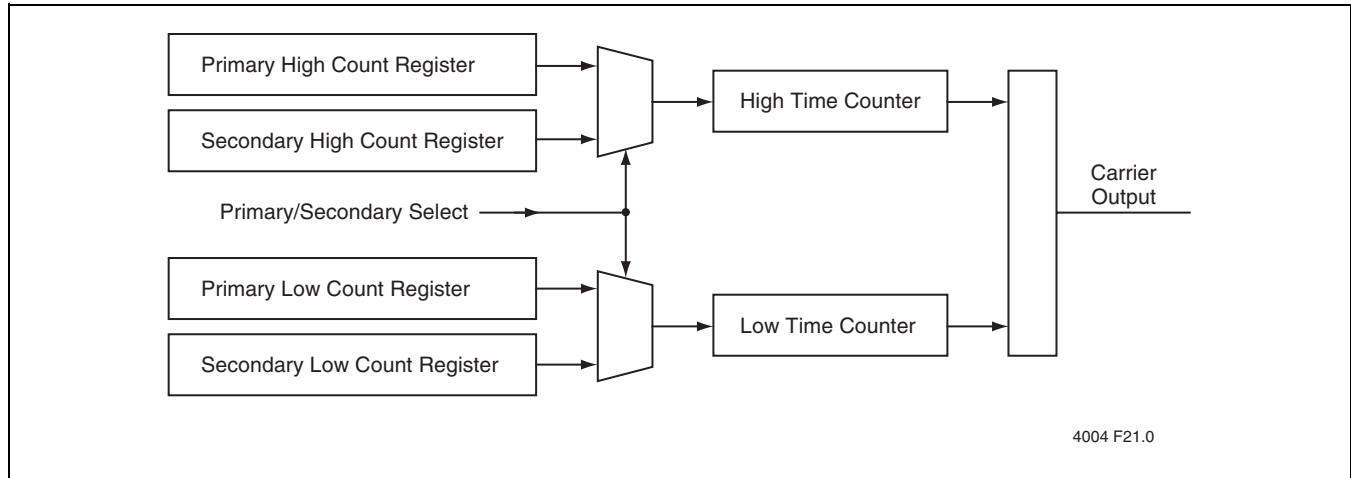


FIGURE 11-2: CARRIER GENERATOR BLOCK DIAGRAM

11.1.1 Time Counter

The high or low time counter is a 6-bit up counter. Only one counter is counting at a certain time. After each increment, the contents of the counter are compared with the appropriate high or low count value register. When the value is reached, the counter is reset and stopped. Meanwhile the other counter is enabled for counting. Assuming that the high time counter is currently active. The carrier output will be high and remain high until it reaches the high time register value. Then, the carrier will be driven low and the low time counter is activated. The carrier output will be low and remain low until it reaches the low time register value. Then, the carrier will be driven high and the high time counter is activated. The cycle repeats automatically generating a periodic carrier signal that is fed into the modulator block. The lowest frequency (maximum period) and highest frequency (minimum period) which can be generated are defined as:

$$f_{\max} = f_{\text{osc}} \div (2 \times 1) \text{ Hz}$$

$$f_{\min} = f_{\text{osc}} \div (2 \times (2^6 - 1)) \text{ Hz}$$

In the general case, the carrier generator output frequency is:

$$f_{\text{out}} = f_{\text{osc}} \div (\text{Highcount} + \text{Lowcount}) \text{ Hz}$$

where: $0 < \text{Highcount} < 64$ and $0 < \text{Lowcount} < 64$

The duty cycle of the carrier signal is:

$$\text{Duty cycle} = \frac{\text{Highcount}}{\text{Highcount} + \text{Lowcount}}$$



11.1.2 Carrier Generator Data Registers (CHR1, CLR1, CHR2, and CLR2)

There are two sets of Carrier Generator Data Registers: Primary and secondary. Each set contains one high time register (CHR1, CHR2) and one low time register (CLR1, CLR2) as shown below.

Carrier Data Register (CHR1)

Address: **0010H**

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	IROLN	CMPOL	PH5	PH4	PH3	PH2	PH1	PH0
Write:								
Reset	0	0	U	U	U	U	U	U

U = Unaffected

Carrier Data Register (CLR1)

Address: **0011H**

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	IROLP	0	PL5	PL4	PL3	PL2	PL1	PL0
Write:								
Reset	0	0	U	U	U	U	U	U

U = Unaffected

Carrier Data Register (CHR2)

Address: **0012H**

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	0	SH5	SH4	SH3	SH2	SH1	SH0
Write:								
Reset	0	0	U	U	U	U	U	U

U = Unaffected

Carrier Data Register (CHR2)

Address: **0013H**

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	0	0	SL5	SL4	SL3	SL2	SL1	SL0
Write:								
Reset	0	0	U	U	U	U	U	U

U = Unaffected

PH0-PH5 and PL0-PL5 Primary Carrier High (PH0-PH5) and Low (PL0-PL5) Time value

These bits contain the number of input clocks for the carrier high and low time periods. When operating in timer mode, this register pair is always selected. When operating in FSK mode, the modulator alternately selects this register pair and the secondary register pair. The primary carrier high and low time values are undefined out of reset. These bits must be written to nonzero values before the carrier generator is enabled to avoid spurious results.

SH0-SH5 and SL0-SL5 Secondary Carrier High (SH0-SH5) and Low (SL0-SL5) Time Value

These bits contain the number of input clocks for the carrier high and low time periods. When operating in time mode, this register pair is never selected. When operating in FSK mode, the modulator alternately selects this register pair and the primary register pair. The secondary carrier high and low



time values are undefined out of reset. These bits must be written to nonzero values before the carrier generator is enabled to avoid spurious results.

CMPOL CMT Output Polarity

This bit controls the polarity of the CMT output (IRO). When this bit is set to zero, the CMT output is active high. When this bit is set to one, the CMT output is active low. The reset state of this bit is zero.

IROLN and IROLP IRO Latch Control

The IRO latch state can be read from either IROLN or IROLP bits. The IRO latch state can be updated with being written on either negative or positive edge of the internal processor clock ($f_{osc}/2$). By writing to IROLN updates the IRO latch on the negative edge. By writing to IROLP updates the IRO latch on the positive edge. The IRO latch is cleared out of reset.

11.2 Modulator

The modulator can operate in two modes (Time and FSK) with a resolution of 4 us (2MHz external oscillator). It can count either system clocks to provide real-time control or it can count carrier clocks for self-clocked protocols. In time mode, the modulator counts clocks derived from the system oscillator and modulates a single-carrier frequency (TIME) or no carrier (baseband). In FSK mode, the modulator counts carrier periods and providing a signal to switch the carrier generator between high/low time register buffers to alternate between two carrier frequencies whenever a modulation period (mark+space counts) expires.

When the modulator is enabled (MCGEN=1), the space period register (SREG) is loaded with the contents of its buffer (SBUFF), the mark buffer register (MBUFF) is loaded into a 12-bit down counter, and the modulator gate is opened for carry signal to pass through. When this counter underflows, the modulator gate is closed and the modulator output is forced to low. The counter is continuously counting down and the logical complement of the contents of the decrementing counter is compared with the SREG. When a match is obtained, the modulator control gate is opened, the MBUFF is re-loaded into the down counter, and SREG is reloaded with the contents of SBUFF. These cycles keep repeating until the modulator is disabled. The current modulator cycle will be allowed to be completed and the modulator output will be forced to low. When SREG=0, the match will happen immediately and no space period will be generated. Some of FSK protocols that require successive bursts of different frequencies need to set SBUFF to 0. The 12-bit MBUFF and SBUFF registers are accessed through three 8-bit modulator period registers MDR1, MDR2, and MDR3. Bit 7 to bit 0 of the down counter can be read from 3FF2. Bit 11 to bit 8 of the down counter can be read from 3FF2 lower 4 bits (upper 4 bits will be 0).

11.2.1 Time mode

When the modulator operates in time mode, the modulation mark and space periods are counted in multiple of ($f_{osc} \div 8$) clocks (=250kHz @ 2MHz osc). This provides a modulator resolution of 4 us. The maximum mark and the maximum space period are 16.384 ms ($2^{12} \times 4$ us). These periods can be doubled by setting the DIV2 bit in the BCSR that will also decrease the resolution to 8us. The modulator control gate and carrier clock are synchronized to prevent modulator output glitches. When the modulator gate is opened (mark), the carrier signal passes through. When modulator gate is closed (space), the modulator output is force to low. If the carrier generator is in the baseband mode (BASE bit in MCSR is set to 1), the modulator output will be force to 1 for the duration of the mark period and force to 0 for the duration of a space period.



Here are the equations to calculate mark and space period for time mode:

$$T_{\text{mark}} = \frac{(\text{MBUFF} + 1) \times 8}{f_{\text{osc}}} \text{sec}$$

$$T_{\text{space}} = \frac{\text{SBUFF} \times 8}{f_{\text{osc}}} \text{sec}$$

Setting the DIV2 bit in the BCSR will double mark and space times.

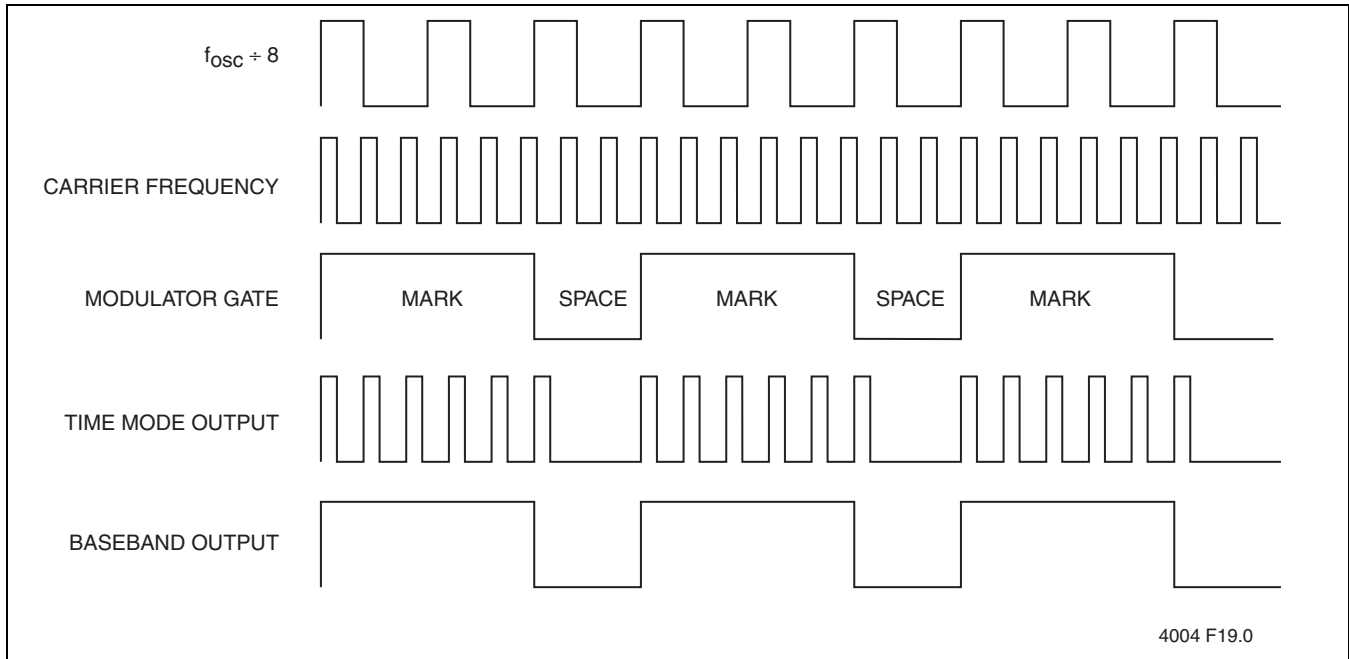


FIGURE 11-3: CMT OPERATION IN TIME MODE

11.2.2 FSK Mode

When the modulator operates in FSK mode, the modulation mark and space periods are counted in multiple of carrier clocks (space period can be zero). The modulator provides a signal to the carrier generator to toggle between primary and secondary data register values whenever the modulator mark period expires. The space period provides an interpulse gap (no carrier), but if SBUFF=0, then the modulator and carrier generator will switch between carrier frequencies without a gap or any carrier glitches (zero space).

Here are equations to calculate mark and space period for FSK mode:

$$T_{\text{mark}} = \frac{\text{MBUFF} + 1}{f_{\text{cg}}} \text{sec}$$

$$T_{\text{space}} = \frac{\text{SBUFF}}{f_{\text{cg}}} \text{sec}$$

Where f_{cg} is the frequency output from the carrier generator. Setting the DIV2 bit in the MCSR will double mark and space times.



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11.2.3 Extended Space Operation

In either time or FSK mode, the space period can be made longer than the maximum possible value of SBUFF. Setting the EXSPC bit in the MCSR will force the modulator to convert the subsequent modulation periods consisting of entirely the space periods with no mark periods. Clearing EXSPC will return the modulator to standard operation at the beginning of the next modulation period.

Here is the equation to calculate the length of an extended space in time mode:

$$T_{\text{espace}} = \frac{((\text{SBUFF1}) + (\text{MBUFF2} + 1 + \text{SBUFF2}) + \dots + (\text{MBUFFn} + 1 + \text{SBUFFn})) \times 8}{f_{\text{osc}}} \text{sec}$$

Where the subscripts 1,2,...n refer to the modulation periods that elapsed while the EXSPC bit was set.

Here is the equation to calculate the length of an extended space in FSK mode:

$$T_{\text{espace}} = \frac{((\text{SBUFF1}) + (\text{MBUFF2} + 1 + \text{SBUFF2}) + \dots + (\text{MBUFFn} + 1 + \text{SBUFFn}))}{f_{\text{cg}}} \text{sec}$$

Where f_{cg} is the carrier frequency output from the carrier generator.

Please note that it is an invalid operation to use extended space (EXSPC=1) at the beginning of a transmission, i.e., do not set extended space (EXSPC) to 1 and change MCGEN from 0 to 1.

11.2.4 End Of Cycle (EOC) Interrupt

At the end of each modulation cycle (when a match of SREG occurs), the end of cycle (EOC) flag is set and an interrupt will be issued to the CPU if the interrupt is enabled (IE=1). Meanwhile, the counter is reloaded from MBUFF. The EOC interrupt provides a means for the user to reload new mark/space values into the MBUFF and SBUFF registers. As the EOC interrupt is coincident with reloading of the counter and SREG, the previous MUBFF and SBUFF contents has been loaded into the counter and SREG respectively. The EOC interrupt service routine (ISR) can update both mark (MBUFF) and space (SBUFF) period values with a new value for the next modulation period. The EOC flag must be cleared within the ISR to prevent another interrupt being generated after exiting the ISR. This EOC flag is cleared by a read of the MCSR follow by an access of MDR2 or MDR3. If the EOC interrupt is not being used (IE=0), the EOC flag need not be cleared.

Modulator Control and Status Register (MCSR)

Address: 0014H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	EOC							
Write:	*	DIV	EIMSK	EXSPC	BASE	MODE	IE	MCGEN
Reset	0	0	0	0	0	0	0	0

* = Unimplemented

Symbol	Function
EOC	End Of Cycle status flag EOC is set when a match occurs between the logical complement of the counter contents and the SBUFF, i.e., the end of the modulator cycle. This flag is cleared by a read of the MCSR follow by an access of MDR2 or MDR3. The EOC flag is cleared by reset. 0: Current modulation cycle in progress 1: End of modulator cycle



DIV2	<p>Divide-by-two prescaler</p> <p>Setting this bit to 1 causes the modulator output to be timed at a twice-slower clock, i.e., doubled the mark and space time. This bit should not be written to during a transmission. This bit is cleared by reset.</p> <p>0: Divide-by-two prescaler disabled 1: Divide-by-two prescaler enabled</p>
EIMSK	<p>External Interrupt Mask</p> <p>The external interrupt mask bit is used to mask IRQ and port B interrupts. This bit is cleared by reset.</p> <p>0: IRQ and port B interrupt enabled 1: IRQ and port B interrupt masked</p>
EXSPC	<p>Extended Space Enable</p> <p>For a description of the extended space enable bit, see Extended Space Operation. This bit is cleared by reset.</p> <p>0: Extended space disabled 1: Extended space enabled</p>
BASE	<p>Baseband Enable</p> <p>Setting this bit to 1 disables the carrier generator and forces the carrier output to high. The modulator output will be force to 1 for the duration of the mark period and force to 0 for the duration of a space period. This bit should not be written to during a transmission. This bit is cleared by reset.</p> <p>0: Baseband disabled 1: Baseband enabled</p>
MODE	<p>Mode Select</p> <p>This bit should not be changed during a transmission operation. This bit is cleared by reset, i.e., default Time mode.</p> <p>0: CMT operates in Time mode. 1: CMT operates in FSK mode.</p>
IE	<p>Interrupt Enabled</p> <p>Setting this bit to 1 will enable interrupt request send to MCU when EOC is set.</p> <p>0: CPU interrupt disabled 1: CPU interrupt enabled</p>
MCGEN	<p>Modulator and Carrier Generator Enable</p> <p>Setting this bit to 1 will enable the carrier generator and modulator. Once enabled, the carrier generator and modulator will function continuously. To prevent spurious operation, the user should initialize all data and control registers before enabling the carrier generator and modulator. When this is set to 0, the current modulator cycle will be allowed to be completed and the modulator output will be forced to low. This bit is cleared by reset.</p> <p>0: Modulator and carrier generator disabled 1: Modulator and carrier generator enabled</p>



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11.2.5 Modulator Period Data Register (MDR1, MDR2, and MDR3)

The MBUFF and SBUFF are 12-bit registers and can be accessed through three 8-bit registers, MDR1, MDR2, and MDR3. MDR2 contains the least significant eight bits of MBUFF (MB7-MB0). MDR3 contains the least significant eight bits of SBUFF (SB7-SB0). MDR1 contains the two most significant nibbles of MBUFF (MB11-MB8) and SBUFF (SB11-SB8).

Modulator Data Register (MDR1)

Address: 0015H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	MB11	MB10	MB9	MB8	SB11	SB10	SB9	SB8
Write:								
Reset	Unaffected by Reset							

Modulator Data Register (MDR2)

Address: 0016H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	MB7	MB6	MB5	MB4	MB3	MB2	MB1	MB0
Write:								
Reset	Unaffected by Reset							

Modulator Data Register (MDR3)

Address: 0017H

	Bit 7	6	5	4	3	2	1	Bit 0
Read:	SB7	SB6	SB5	SB4	SB3	SB2	SB1	SB0
Write:								
Reset	Unaffected by Reset							

11.2.5.1 Transmitter Block

The state of infrared output pin IRO is controlled by the transmitter output block.

When the modulator/carrier generator is enabled, the IRO pin state is gated by the modulator output. Otherwise, the IRO pin is controlled by the state of the IRO latch, which is described in CHR1 and CLR1 bit 7.

11.2.5.2 Idle Mode Operation

During idle mode, if the CMT was enabled, it continues to operate normally. The CMT will not be able to be updated because the MCU is in idle mode and not operating. If there has one pending CMT interrupt when idle mode is entered, the pending interrupt will be served and pull the chip out of idle mode. Pending or new CMT interrupt will bring the chip out of idle mode.

11.2.5.3 Stop Mode Operation

During stop mode, the CMT halts all operation and no registers are affected.



12.0 PROGRAMMING FLOW DIAGRAM

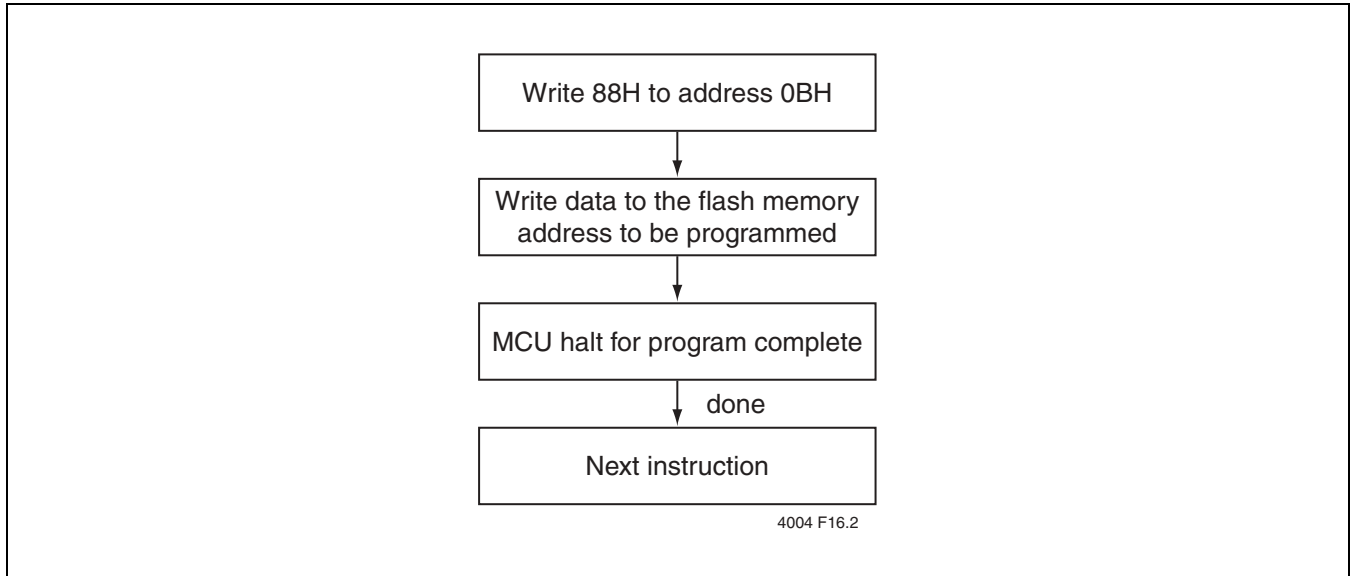


FIGURE 12-1: IN-APPLICATION PROGRAMMING

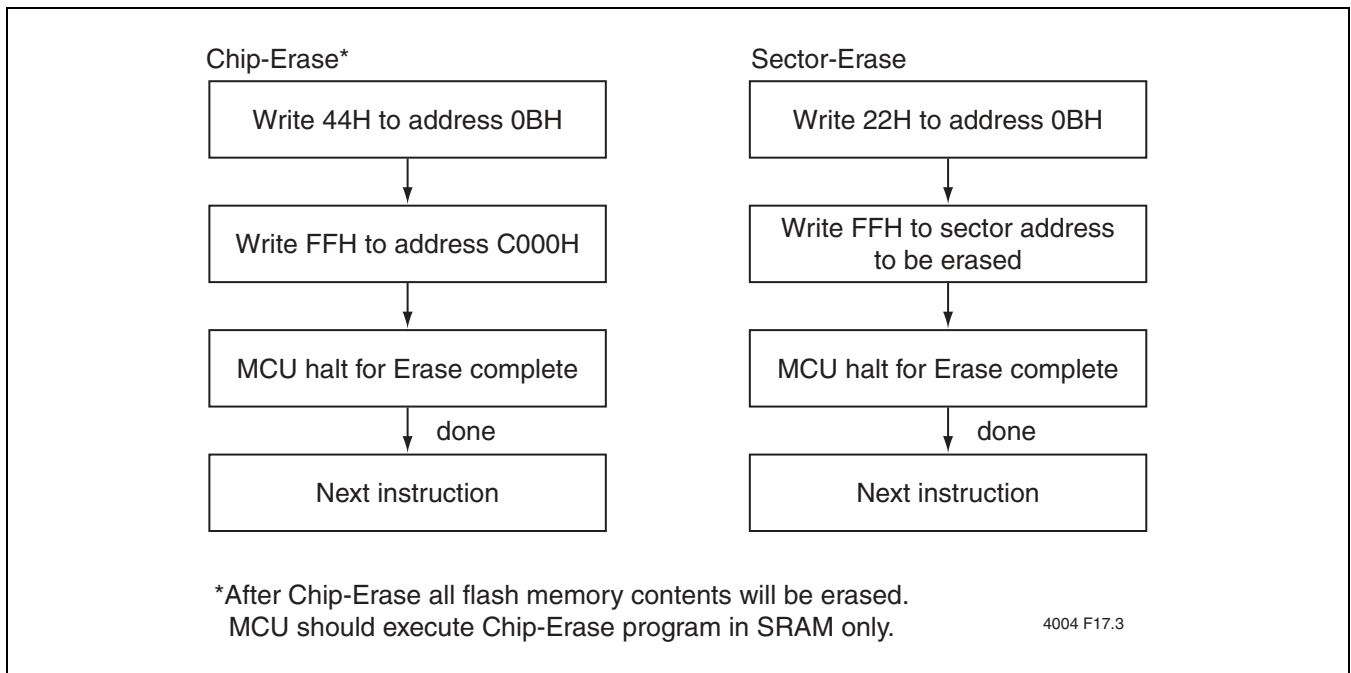


FIGURE 12-2: IN-APPLICATION ERASE

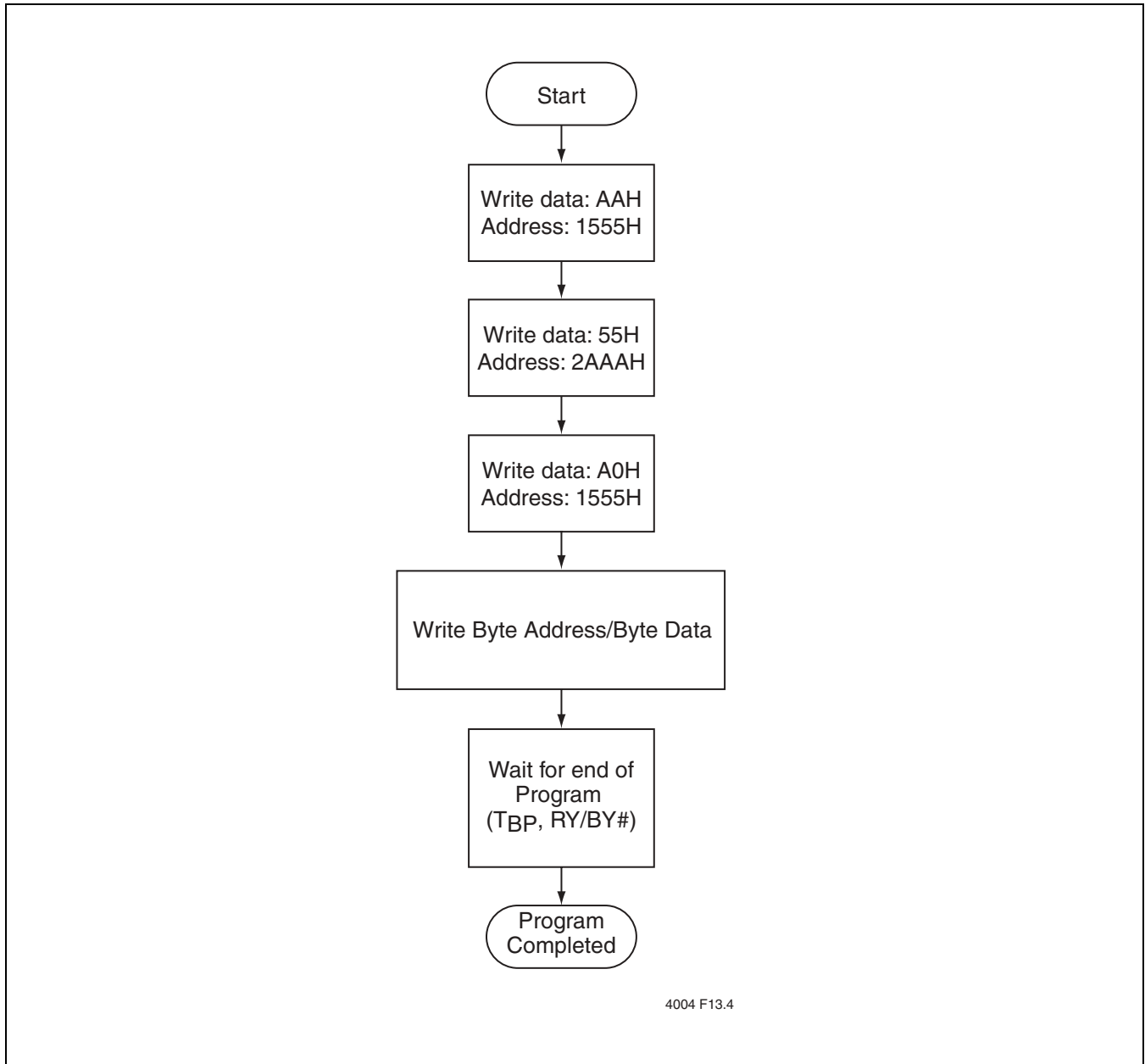
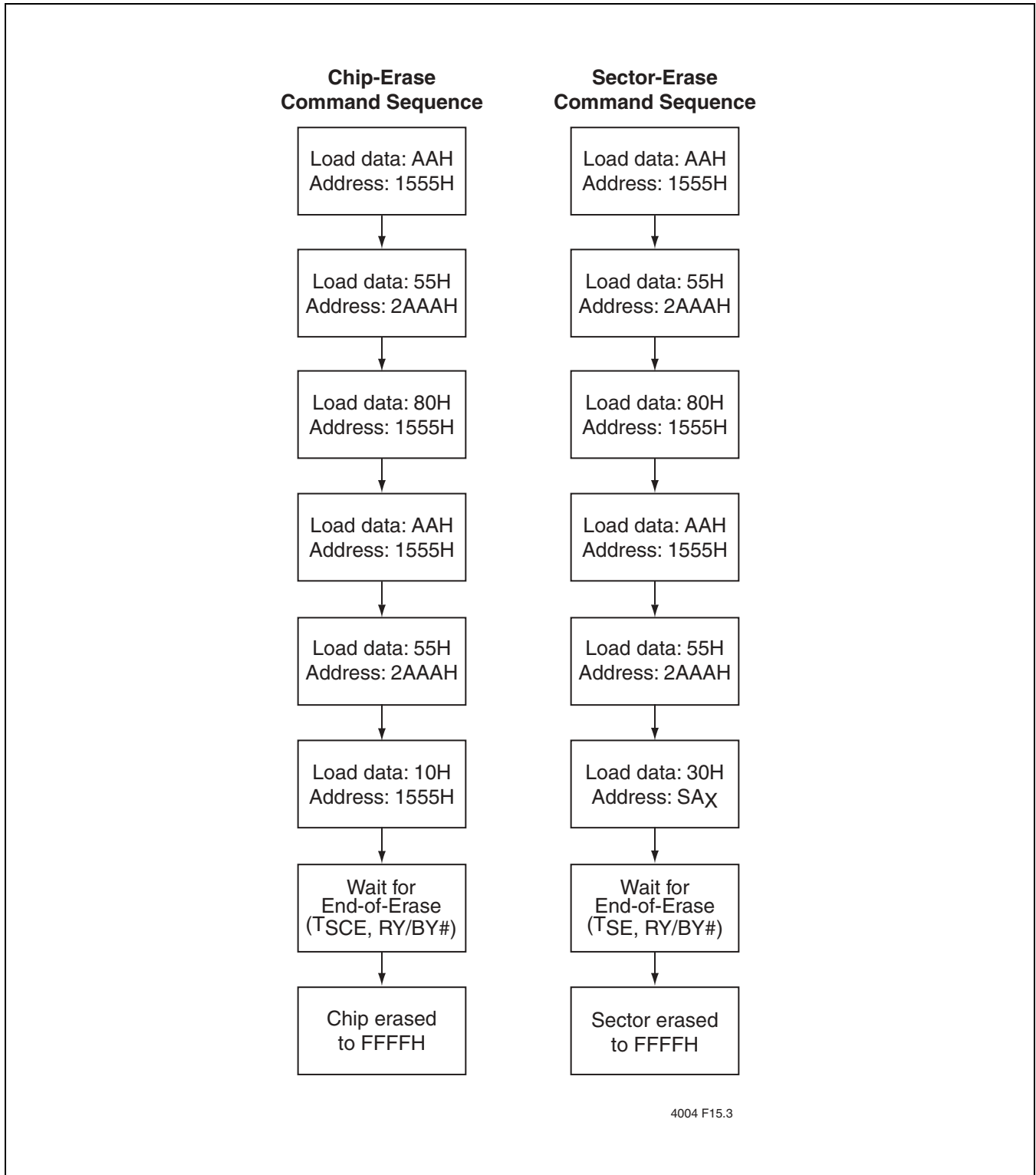


FIGURE 12-3: EXTERNAL FLASH BYTE-PROGRAM ALGORITHM FOR EXTERNAL FLASH PROGRAMMING MODE

Note: Please refer to the SST65P542R data sheet for more information.



4004 F15.3

FIGURE 12-4: CHIP-/SECTOR-ERASE COMMAND SEQUENCE FOR EXTERNAL FLASH PROGRAMMING MODE

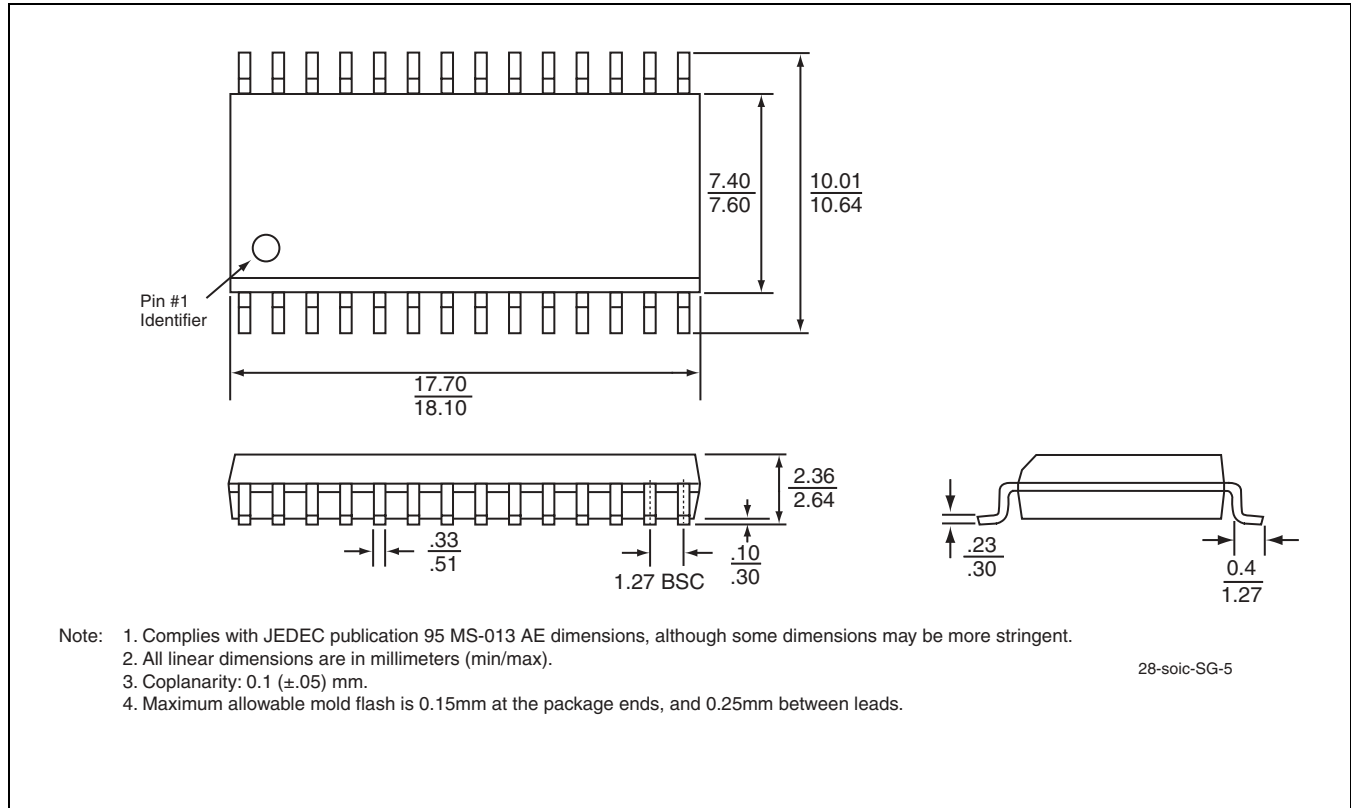
Note: Please refer to the SST65P542R data sheet for more information.



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13.0 PACKAGING DIAGRAMS



28-PIN SMALL OUTLINE INTEGRATED CIRCUIT (SOIC)
SST PACKAGE CODE: SG